

# Stubborn and Dead-Zone Redesign for State Observers and Dynamic Output Feedback

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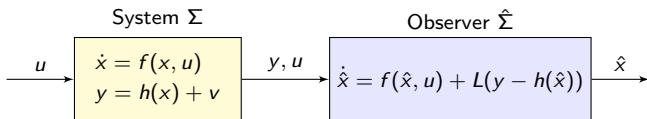
# Outline

- 1 Introduction
- 2 Observer Class
- 3 Stubborn Redesign
- 4 Dead-Zone Redesign
- 5 Synchronization
- 6 Dynamic Output Feedback
- 7 Conclusions

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# The observation problem



## System $\Sigma$

- state  $x \in \mathbb{R}^n$
- known external input  $u \in U$
- measured output  $y \in \mathbb{R}^m$

## Observer $\hat{\Sigma}$

- estimate  $\hat{x} \in \mathbb{R}^n$
- unknown measurement noise  $v \in \mathbb{R}^m$

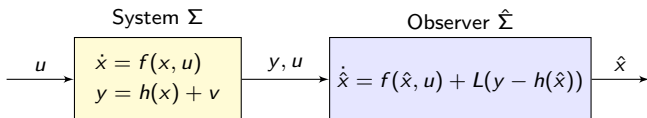
## ISS Observer

The observer  $\hat{\Sigma}$  is ISS (input-to-state stable) if

$$|x(t) - \hat{x}(t)| \leq \beta(|x(0) - \hat{x}(0)|, t) + \gamma \left( \sup_{s \in [0, t]} |v(s)| \right)$$

for all  $t \geq 0$ , for some  $\beta \in \mathcal{KL}$  and  $\gamma \in \mathcal{K}$ .

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# Properties of an observer

## Characteristics of an ISS Observer

- $\beta$  characterizes the performance in nominal conditions ( $v = 0$ ).
- $\gamma$  characterizes the robustness with respect to  $v$ .

$$|x(t) - \hat{x}(t)| \leq \beta(|x(0) - \hat{x}(0)|, t) + \gamma \left( \sup_{s \in [0, t]} |v(s)| \right)$$

- Ideal behavior of an ISS observer:

✓ Fast convergence:

for example,  $\beta(s, t) = ae^{-bt}|s|$ ,  $a, b > 0$ , with  $b$  “large.”

✓ Small Peaking:

for example,  $\beta(s, t) = ae^{-bt}|s|$ ,  $a, b > 0$ , with  $a$  “small.”

✓ Small asymptotic gain:

for example,  $\gamma(s) = \bar{\gamma}|s|$ , with  $\bar{\gamma} > 0$  “small.”

✗ Trade-off between speed of convergence and asymptotic gain

# The linear case: an example

Consider a linear system and a linear (Luenberger) observer

$$\Sigma : \begin{cases} \dot{x} = Ax + Bu \\ y = Cx + v \end{cases} \quad \widehat{\Sigma} : \dot{\hat{x}} = A\hat{x} + Bu + \mathbf{L}(y - C\hat{x})$$

with

- $(A, C)$  detectable pair
- $\mathbf{L}$  gain of the observer to be chosen so that  $A - \mathbf{L}C$  is Hurwitz.

The dynamics of the estimation error  $\tilde{x} := x - \hat{x}$  is given by

$$\dot{\tilde{x}} = (A - \mathbf{L}C)\tilde{x} - \mathbf{L}v$$

and thus

$$|x(t) - \hat{x}(t)| \leq \left| e^{(A - \mathbf{L}C)t} \right| |x(0) - \hat{x}(0)| + \gamma(|v|_\infty), \quad \begin{cases} (s, t) \mapsto \beta(s, t) := |e^{(A - \mathbf{L}C)t}| |s| \\ s \mapsto \gamma(s) := |s| |\mathbf{L}| \int_0^\infty |e^{(A - \mathbf{L}C)\tau}| d\tau \end{cases}$$

(finite-gain exponentially ISS).

- ✗ We cannot make both  $A - \mathbf{L}C$  s.t.  $|e^{(A - \mathbf{L}C)t}| \leq M \exp(-\alpha t)$  with  $\alpha$  large (i.e., fast transient) and  $\gamma(\cdot)$  small (i.e. insensitive to noise)

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# Observer gain design for linear systems

- Based on additional hypothesis on  $v$ , we can figure out different approaches:
  - $H_\infty$  design allows to minimize  $\gamma$  over all frequencies.
  - If  $v$  acts on some known frequencies  $\omega \in [\underline{\omega}, \bar{\omega}]$ , minimize the gain

$$G(j\omega) = [j\omega I - (A - LC)]^{-1}L.$$

- Kalman filter for optimal gain design
- ...

# Observer design for nonlinear systems

- ✓ Exploring some structural properties of the system, many different designs exist

Property	Observer Technique
Detectability	Kazantis-Kravaris Luenberger (KKL) observers
Uniform observability	High-gain (HGO) observers
Lipschitz systems	LMI or circle-criterion approach
Input-affine systems	Riccati-like approach
Local observability	extended Kalman filters
...	...

- ✗ [Shim, Seo, & Teel, *Automatica* 2003, p. 890] and reference therein pointed out the fragility (lack of ISS) of certain nonlinear observers
- ✗ Few tools to analyze the effect of noises in the nonlinear framework:
  - ISS gains based on Lyapunov analysis [Alessandri, *Mathematics* 2020];
  - analysis of measurement noise in high-gain observers [Sanfelice & Praly, *Automatica* 2011], [Astolfi, Marconi, Praly, & Teel, *NOLCOS* 2016].
- ✗ Techniques to improve sensitivity to measurement noise are developed *ad hoc* (i.e. for specific classes of systems and/or observers)

# Plan of the talk

- We follow a **redesign approach**.
- We focus on two special classes of measurement-noise perturbations  $v$ :
  - 1) outliers (i.e. sporadic impulsive noise);
  - 2) persistent “small” noise.
- Two techniques will be developed:
  - 1) dynamic saturation redesign (“stubborn redesign”);
  - 2) dynamic dead-zone redesign.
- We will provide sufficient conditions to apply a general paradigm for the purpose of redesign of
  - 1) state observers for linear and nonlinear systems;
  - 2) static output feedback for synchronization of multi-agent systems;
  - 3) dynamic output feedback of linear plants.

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  - 3) dynamic output feedback of linear plants.

# Bibliography

- The first part of this presentation is based on the following literature:
  - Stubborn redesign for linear observers
    - [Alessandri, Zaccarian, "Results on stubborn Luenberger observers for linear time-invariant plants," ECC 2015]
    - [Alessandri, Zaccarian, "Stubborn state observers for linear time-invariant systems," Automatica 2018]
  - Stubborn redesign for high-gain observers
    - [Astolfi, Alessandri, Zaccarian, "Stubborn ISS redesign for nonlinear high-gain observers," IFAC WC 2017]
  - Dead-zone redesign for linear observers
    - [Cocetti, Tarbouriech, Zaccarian, "On dead-zone observers for linear plants," ACC 2018]
  - Dead-zone redesign for high-gain observers
    - [Cocetti, Tarbouriech, Zaccarian, "High-gain dead-zone observers for linear and nonlinear plants," IEEE LCSS 2019]
  - Stubborn and dead-zone redesign for nonlinear estimators
    - [Astolfi, Alessandri, Zaccarian, "Stubborn and dead-Zone redesign for nonlinear observers and filters," IEEE TAC 2021]

# Bibliography

- The second part of this presentation is adapted from

[Casadei, Astolfi, Alessandri, Zaccarian, "Synchronization of interconnected linear systems via dynamic saturation redesign," IFAC NOLCOS 2019]

[Casadei, Astolfi, Alessandri, Zaccarian, "Synchronization in networks of identical nonlinear systems via dynamic dead zones," IEEE LCSS 2019]

- The third part of this presentation is based on

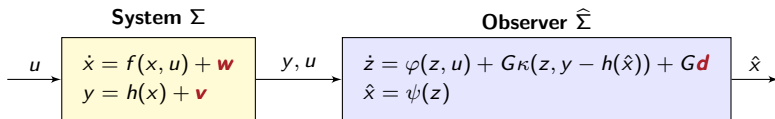
[Tarbouriech, Alessandri, Astolfi, Zaccarian, "LMI-based stubborn and dead-zone redesign in linear dynamic output feedback," 61st CDC and IEEE LCSS 2022]

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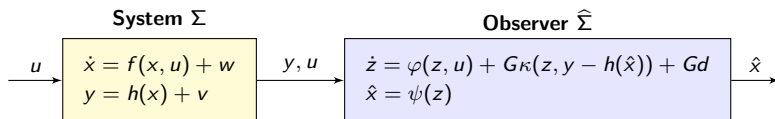
# Main Assumptions



- state  $x \in \mathbb{R}^n$
- known external input  $u \in U$
- measured output  $y \in \mathbb{R}^m$
- unknown perturbation  $\mathbf{w} \in W$
- observer state  $z \in \mathbb{R}^o, o \geq n$
- state estimate  $\hat{x} \in \mathbb{R}^n$
- unknown measurement noise  $\mathbf{v} \in V$
- unknown perturbation  $\mathbf{d} \in D$
- $G$  is a selection matrix
- $\kappa$  is the correction term
- $\psi$  maps the state of the observer  $z$  in the actual estimate  $\hat{x}$
- We will use  $D^+$  to denote the (upper-right) Dini derivative

$$D^+ V(t) := \limsup_{h \rightarrow 0} \frac{V(t+h) - V(t)}{h}$$

# Main Assumptions



## Assumption 1 (ISS Observer with ISS Lyapunov function)

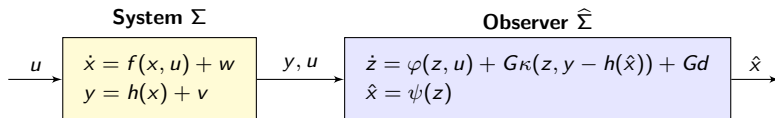
There exists  $V : X \times Z \rightarrow \mathbb{R}_{\geq 0}$ , a function  $\psi^{-R} : X \rightarrow Z$ ,  $\underline{\alpha}, \bar{\alpha} \in \mathcal{K}_{\infty}$ , and  $c, \bar{\kappa}, c_v, c_w > 0$  such that

- 1  $x = \psi(\psi^{-R}(x))$  (pseudo-right-inverse)
- 2  $\|G\| \leq 1, \quad |\kappa(z, y_1) - \kappa(z, y_2)| \leq \bar{\kappa}|y_1 - y_2|$  (Lipschizianity)
- 3  $\underline{\alpha}(|x - \psi(z)|) \leq V(x, z) \leq \bar{\alpha}(|\psi^{-R}(x) - z|)$  ("sandwich")
- 4  $D^+V \leq -cV(x, z) + c_v|v| + c_w|w| + c_d|d|$  ("ISS bound")

for all  $x \in X, u \in U, z \in Z, y_1, y_2 \in \mathbb{R}^m, (v, w, d) \in V \times W \times D$ .

- The observer is supposed to be ISS also with respect to system disturbances  $w$  and observer perturbations  $d$ .

# Main Assumptions



## Assumption 2 (output-growth condition)

There exists  $\ell_0, \ell_1, \ell_v, \ell_w, \ell_d > 0$  such that

$$\text{5 } |h(x) - h(\hat{x})| \leq \ell_0 V(x, z)$$

$$\text{6 } |D^+(h(x) - h(\hat{x}))| \leq \ell_1 V(x, z) + \ell_v |v| + \ell_w |w| + \ell_d |d|$$

for all  $x \in X$ ,  $u \in U$ ,  $z \in Z$ ,  $y_1, y_2 \in \mathbb{R}^m$ ,  $(v, w, d) \in V \times W \times D$ .

- Recall that  $V(x, z) \geq \alpha(|x - \psi(z)|)$ . Hence condition 5 holds if

$$|h(x) - h(\hat{x})| = |h(x) - h(\psi(z))| \leq k_0 \alpha(|x - \psi(z)|), \quad k_0 > 0.$$

- Condition 6 imposes a growth on the derivative of  $y - \hat{y}$ , with  $\hat{y} := h(\hat{x})$ .

## Example: Input-affine systems (Besançon et al, 1996)

$$\Sigma : \begin{cases} \dot{x} = A(u)x + Bu \\ y = Cx \end{cases} \quad \hat{\Sigma} : \begin{cases} \dot{\hat{x}} = A(u)\hat{x} + Bu + P^{-1}C^T(y - C\hat{x}) \\ \dot{P} = -2\mu P - A(u)^T P - PA(u) + 2C^T C \end{cases}$$

Observer with Lipschitz output injection term

$$\dot{z} = \varphi(z, u) + G\kappa(z, y - h(\hat{x})), \quad \hat{x} = \psi(z)$$

$$1 \quad x = \psi(\psi^{-R}(x))$$

$$2 \quad |G| \leq 1, \quad |\kappa(z, y_1) - \kappa(z, y_2)| \leq \bar{\kappa}|y_1 - y_2|$$

$$\checkmark \quad z = (\hat{x}, \text{vec}(P)), \quad \psi(z) = [I \ 0](\hat{x}, \text{vec}(P)) = \hat{x}, \quad \psi^{-R}(x) = (x, 0)$$

$$\checkmark \quad G = [I; 0], \quad \kappa(z, s) = P^{-1}C^T s$$

$$\checkmark \quad \text{Lipschitzianity of } \kappa \text{ follows from boundedness of } P.$$

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✓  $z = (\hat{x}, \text{vec}(P)), \psi(z) = [I \ 0](\hat{x}, \text{vec}(P)) = \hat{x}, \psi^{-R}(x) = (x, 0)$

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✓ Lipschitzianity of  $\kappa$  follows from boundedness of  $P$ .

# Example: Input-affine systems (Sandwich and Directional Derivative)

$$\Sigma : \begin{cases} \dot{x} = A(u)x + Bu + \mathbf{w} \\ y = Cx + \mathbf{v} \end{cases} \quad \hat{\Sigma} : \begin{cases} \dot{\hat{x}} = A(u)\hat{x} + Bu + P^{-1}C^T(y - C\hat{x}) + \mathbf{d} \\ \dot{P} = -2\mu P - A(u)^T P - PA(u) + 2C^T C \end{cases}$$

## Observer with ISS Lyapunov function

$$3 \quad \underline{\alpha}(|x - \psi(z)|) \leq \mathbf{V}(x, z) \leq \bar{\alpha}(|\psi^{-R}(x) - z|)$$

$$4 \quad D^+ \mathbf{V} \leq -c \mathbf{V}(x, z) + c_v |\mathbf{v}| + c_w |\mathbf{w}| + c_d |\mathbf{d}|$$

- Suppose PE is verified  $\implies \underline{p}I \leq P(t) \leq \bar{p}I \quad \forall t \geq 0$ .
- Select  $\mathbf{V}(x, z) := \sqrt{W(z, x)}$ ,  $W(z, x) := (x - \hat{x})^T P(x - \hat{x})$ .

$$\checkmark \quad \sqrt{\underline{p}}|x - \hat{x}| \leq \mathbf{V}(z, x) \leq \sqrt{\bar{p}}|x - \hat{x}| \leq \sqrt{\bar{p}}|\psi^{-R}(x) - z|, \quad \begin{cases} \psi^{-R}(x) = (x, 0) \\ z = (x, \text{vec}(P)) \end{cases}$$

✓ Furthermore,  $\dot{W} = -2\mu W + 2(x - \hat{x})^T [P(\mathbf{w} - \mathbf{d}) + C^T \mathbf{v}]$ , which gives

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# Example: Input-affine systems (Output Growth Condition)

$$\Sigma: \begin{cases} \dot{x} = A(u)x + Bu + \mathbf{w} \\ y = Cx + \mathbf{v} \end{cases} \quad \hat{\Sigma}: \begin{cases} \dot{\hat{x}} = A(u)\hat{x} + Bu + P^{-1}C^{\top}(y - C\hat{x}) + \mathbf{d} \\ \dot{P} = -2\mu P - A(u)^{\top}P - PA(u) + 2C^{\top}C \end{cases}$$

Observer with ISS Lyapunov function and output growth-condition

$$5 \quad |h(x) - h(\hat{x})| \leq \ell_0 \mathbf{V}(x, z)$$

$$6 \quad |D^+(h(x) - h(\hat{x}))| \leq \ell_1 \mathbf{V}(x, z) + \ell_v |\mathbf{v}| + \ell_w |\mathbf{w}| + \ell_d |\mathbf{d}|$$

$$\checkmark \quad |C(x - \hat{x})| \leq \frac{|C|}{\sqrt{p}} \mathbf{V}(x, z)$$

$$\begin{aligned} \checkmark \quad |D^+ C(x - \hat{x})| &\leq |C(A(u) - P^{-1}C^{\top}C)(x - \hat{x}) - CP^{-1}C^{\top}\mathbf{v} + C\mathbf{d} + C\mathbf{w}| \\ &\leq \underbrace{(|C| \sup |A(u)| + |C|^3 \underline{p}^{-1})(\sqrt{p})^{-1} \mathbf{V}(x, z)}_{\ell_1} + \underbrace{\underline{p}^{-1}|C|^2}_{\ell_v} |\mathbf{v}| + \underbrace{|C|}_{\ell_w} |\mathbf{w}| + \underbrace{|C|}_{\ell_d} |\mathbf{d}| \end{aligned}$$

## Other examples

Assumptions 1-2 are verified by the following observer techniques:

[Astolfi, Alessandri & Zaccarian (2021)]

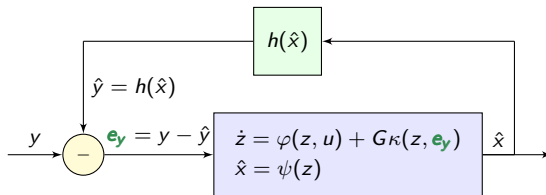
- Kalman filter and extended Kalman filter [Kalman (1960), others]
- Linear Luenberger observers [Luenberger (1971)]
- Observers for Lipschitz systems based on LMI design or circle criterion  
[Rajamani (1998), Arcak & Kokotovic (2001)]  
[Zemouche & Boutayeb (2013)]
- Observers for input-affine systems with Riccati design  
[Besançon & Bornard & Hammouri (1996)]
- High-gain observers [Tornambé (1991), Khalil (1992)]  
[Gauthier & Kupka (2001)]
- Kazantzis-Kravaris/Luenberger observer [Andrieu & Praly (2006)]
- ...

# Outline

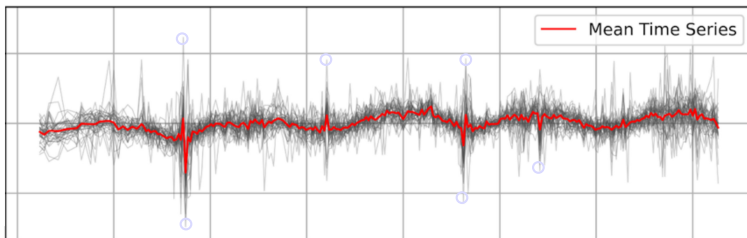
- 1 Introduction
- 2 Observer Class
- 3 Stubborn Redesign**
- 4 Dead-Zone Redesign
- 5 Synchronization
- 6 Dynamic Output Feedback
- 7 Conclusions

# How to deal with outliers?

Observer:

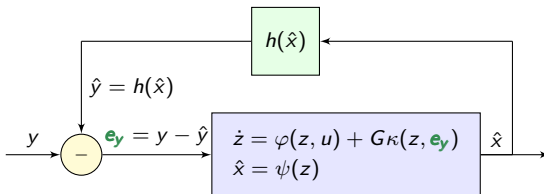


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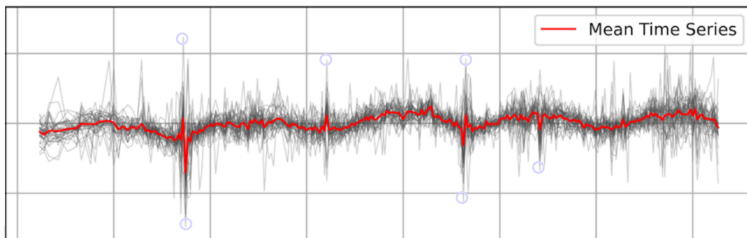


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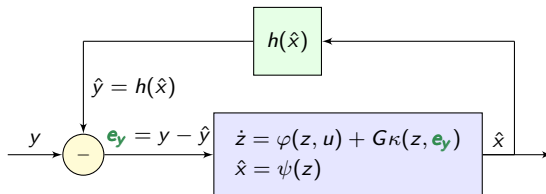


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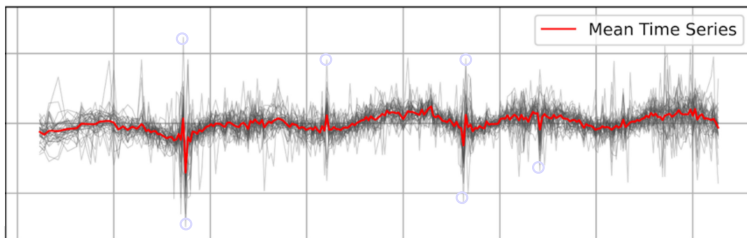


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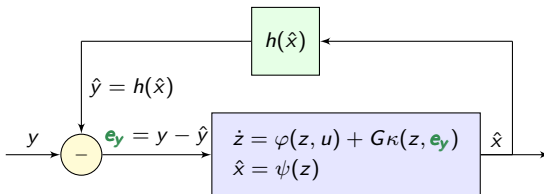


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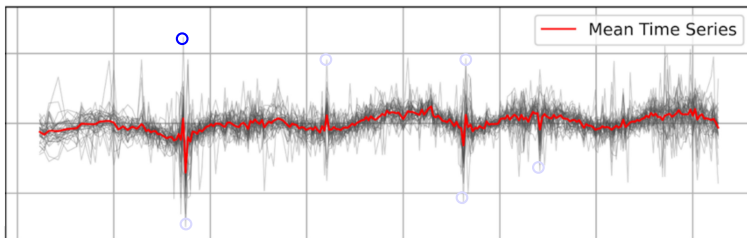


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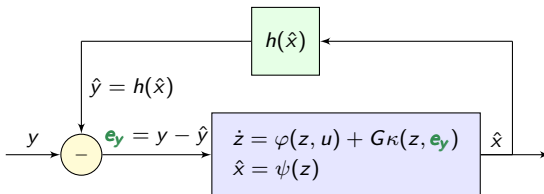


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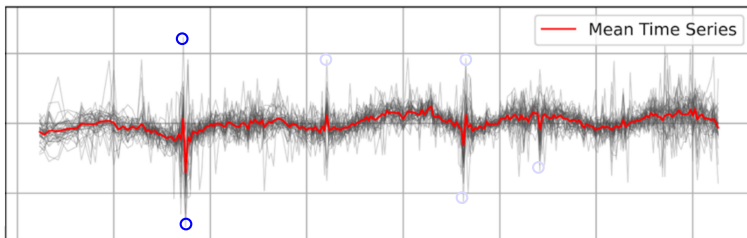


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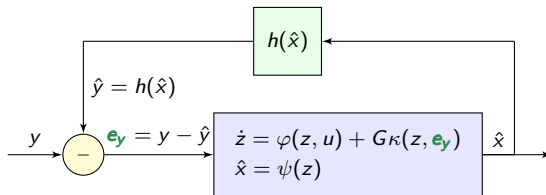
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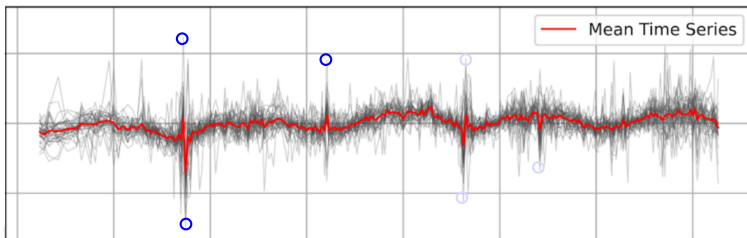


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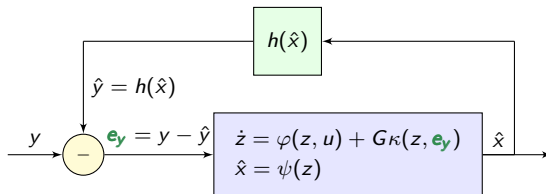


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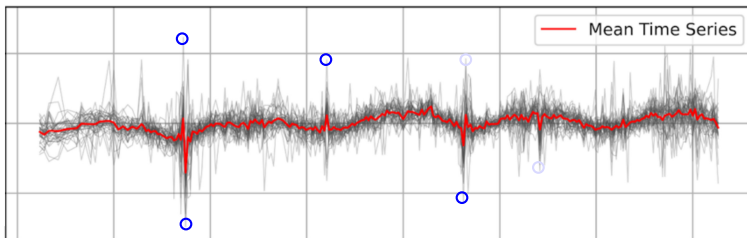


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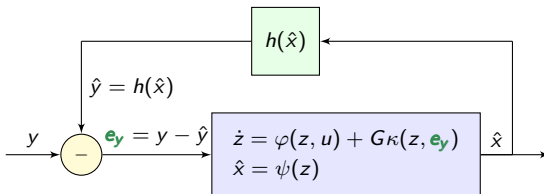


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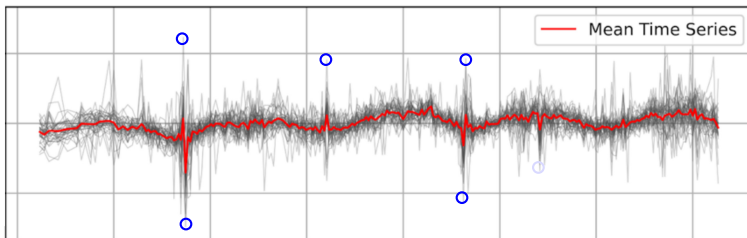


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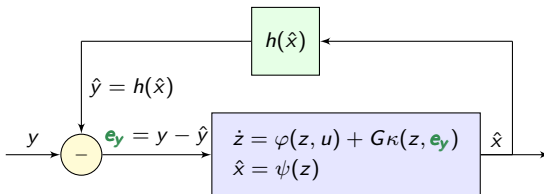


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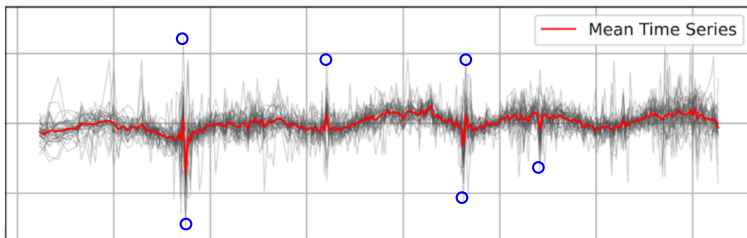


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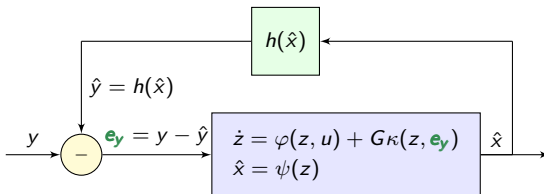


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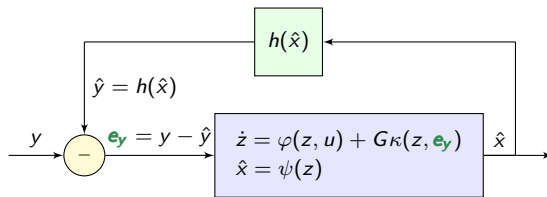
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**Main idea:**

# How to deal with outliers?

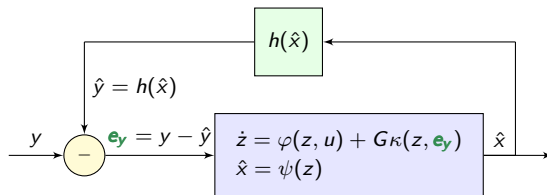
Observer:



- Suppose  $\mathbf{e}_y$  is very small for a long amount of time  
 $\Rightarrow x$  and  $\hat{x}$  are close to each other  
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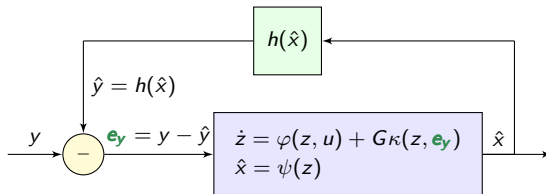
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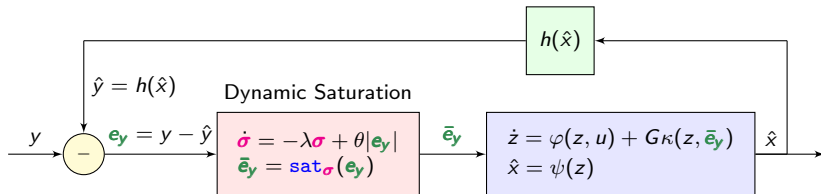


- Suppose  $\mathbf{e}_y$  is very small for a long amount of time
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  - $\Rightarrow$  we **don't want to** use the correction term  $\kappa(z, \mathbf{e}_y)$
- Suppose  $\mathbf{e}_y$  is large but for a very short amount of time
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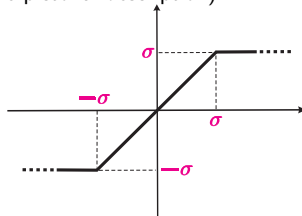


# A Dynamic Saturation Scheme

We modify the previous structure by adding a dynamic saturation for  $\mathbf{e}_y = y - \hat{y}$ , i.e.,



where  $\text{sat}_\sigma(s) := \max\{-\sigma, \min\{\sigma, s\}\}$  and  $\lambda, \theta > 0$  (from now on we consider a scalar output to simplify the pictorial description).

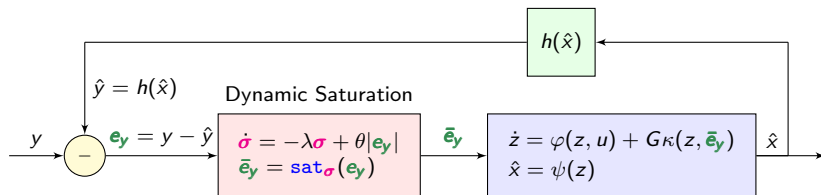


What happens in the presence of an outlier?

- If  $\mathbf{e}_y$  is persistently small then  $\sigma$  becomes small.
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- The level of the saturation is selected as the current value of  $\sigma$ .
- The level of  $\sigma$  is dynamically adapted based on the norm of  $\mathbf{e}_y = y - \hat{y}$ .

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- If an outlier occurs,  $\mathbf{e}_y$  becomes large but  $\bar{\mathbf{e}}_y$  is saturated at the current (small) value of  $\sigma$ , thus mitigating the outlier's effect on the estimate.
- If  $\mathbf{e}_y$  is persistently large then  $\sigma$  becomes large and we desaturate (avoids intrinsic limitations of saturated feedback).

# Saturation Redesign: Main Result

## Theorem 1 [Astolfi, Alessandri & Zaccarian, IEEE TAC 2021]

Consider a given system  $\Sigma$  and observer  $\hat{\Sigma}$

$$\Sigma : \begin{cases} \dot{x} = f(x, u) + w \\ y = h(x) + v \end{cases} \quad \hat{\Sigma} : \begin{cases} \dot{z} = \varphi(z, u) + G\kappa(z, y - h(\hat{x})) + Gd \\ \hat{x} = \psi(z) \end{cases}$$

Suppose that  $\hat{\Sigma}$  is an ISS Observer for  $\Sigma$  satisfying the output-growth condition. Then, for any  $\lambda > 0$  there exists a  $\theta^* > 0$  such that, for any  $\theta > \theta^*$ , the observer

$$\hat{\Sigma}_{\text{sat}} : \begin{cases} \dot{z} = \varphi(z, u) + G\kappa(z, \text{sat}_{\sigma}(y - h(\hat{x}))) + Gd \\ \dot{\sigma} = -\lambda\sigma + \theta|y - h(\hat{x})| \\ \hat{x} = \psi(z) \end{cases}$$

is an ISS Observer for  $\Sigma$  with ISS Lyapunov function satisfying the output-growth condition.

- If  $y \in \mathbb{R}^m$ ,  $m \geq 1$ , then  $\hat{\Sigma}_{\text{sat}}$  reads

$$\begin{aligned} \dot{z} &= \varphi(z, u) + G\kappa\left(z, [\text{sat}_{\sigma_1}(y_1 - h_1(\hat{x})), \dots, \text{sat}_{\sigma_m}(y_m - h_m(\hat{x}))]\right) \\ \dot{\sigma}_i &= -\lambda_i\sigma_i + \theta_i|y_i - h_i(\hat{x})| \quad i = 1, \dots, m, \end{aligned}$$

- In general,  $\lambda \leq \theta$

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# Sketch of Proof of Theorem 1

- Rewrite the observer  $\widehat{\Sigma}_{\text{sat}}$  as

$$\begin{cases} \dot{\sigma} = -\lambda\sigma + \theta|y - h(\hat{x})| \\ \dot{z} = \varphi(z, u) + G\kappa(z, y - h(\hat{x})) + G\delta \\ \delta = \kappa(z, \text{sat}_{\sigma}(y - h(\hat{x}))) - \kappa(z, y - h(\hat{x})). \end{cases}$$

- In light of the Lipschitz properties of  $\kappa$  and saturation function, we have

$$|\delta| \leq \bar{\kappa} |\text{sat}_{\sigma}(y - h(\hat{x})) - (y - h(\hat{x}))| \leq \bar{\kappa} |y - h(\hat{x})|.$$

- Use the Lyapunov function

$$(x, z, \sigma) \mapsto W(x, z, \sigma) = V(x, z) + \zeta\sigma + (\zeta + \eta) \max\{|y - h(\hat{x})| - \sigma, 0\}$$

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- When  $|y - h(\hat{x})| \leq \sigma$ , then  $\delta = 0$ , and we have

$$D^+W \leq -cV(x, z) - \zeta\lambda\sigma + \zeta\theta|y - h(\hat{x})|$$

using the output-growth condition we obtain

$$D^+W \leq -(c - \zeta\theta\ell_0)V(x, z) - \zeta\lambda\sigma \leq 0$$

for  $\zeta$  small enough.

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$$D^+W \leq -cV(x, z) + c_d|\delta| + (\zeta + \eta)|D^+(y - h(\hat{x}))| + \eta\lambda\sigma - \eta\theta|y - h(\hat{x})|.$$

By using the output-growth condition we obtain

$$\begin{aligned} D^+W &\leq -[c - (\zeta + \eta)(\ell_1 + \ell_0\bar{\kappa}\ell_d)]V(x, z) - (\theta\eta - \lambda\eta + c_d\bar{\kappa}\ell_0)|y - h(\hat{x})| \\ &\leq 0 \end{aligned}$$

for  $\zeta, \eta$  small enough and  $\theta$  large enough.

# Sketch of Proof of Theorem 1

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with  $\zeta, \eta > 0$ .

- ✓ We conclude that  $D^+W \leq -\varepsilon W$  for some  $\varepsilon > 0$ .
- ✓ The analysis can be done with  $w, v, d$  to show the desired ISS properties.
- ✓ It is not too hard to verify also the other properties for the redesigned observer.



# Noise effect analysis for linear systems

- Consider the linear case

$$\Sigma : \begin{cases} \dot{x} = Ax + Bu \\ y = Cx + \mathbf{v} \in \mathbb{R} \end{cases} \quad \widehat{\Sigma} : \begin{cases} \dot{\hat{x}} = A\hat{x} + Bu + \mathbf{L}(y - C\hat{x}) \end{cases}$$

and the (nonlinear) redesigned observer

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- Consider the error variables

$$\begin{aligned} \tilde{x}_0 &:= x - \hat{x}, & \hat{x} &\in \widehat{\Sigma}, \\ \tilde{x}_{\text{sat}} &:= x - \hat{x}, & \hat{x} &\in \widehat{\Sigma}_{\text{sat}}. \end{aligned}$$

- We are interested in analyzing the effect of two types of measurement noise  $\mathbf{v}$ :
  - impulsive noise (outlier);
  - constant noise.

# Effect of outliers in the linear case

Proposition 1 [Astolfi, Alessandri & Zaccarian, IEEE TAC 2021]

- Suppose  $\mathbf{v}$  is a piecewise constant perturbation of the form

$$\mathbf{v}(t) = \delta_{\tau}(t) = \begin{cases} \frac{1}{\tau} & 0 \leq t \leq \tau \\ 0 & t > \tau. \end{cases}$$

- Suppose  $A$  is non-singular.
- Consider the solutions of  $\hat{\Sigma}$  and  $\hat{\Sigma}_{\text{sat}}$  with  $\tilde{x}_0(0) = \tilde{x}_{\text{sat}}(0) = 0$  and  $\sigma(0) = 0$ .

Then, as  $\tau$  tends to  $0^+$ , we have

$$|\tilde{x}_{\text{sat}}(\tau)| \leq 2\tau\theta|\tilde{x}_0(\tau)|.$$

Furthermore,  $|\tilde{x}_{\text{sat}}(t)|$  converges to zero if  $\tau \rightarrow 0^+$ .

- ✓ The performance in the presence of outliers is improved.
- ✓ If the outlier is instantaneous ( $\tau = 0$ ), its effect is completely erased!
- After  $t \geq \tau$ , there are no more perturbations  $\mathbf{v} = 0$   
 $\implies$  the observers evolves with initial conditions such that  $|\tilde{x}_{\text{sat}}(0)| < |\tilde{x}_0(0)|$ .

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# Effect of constant noise in the linear case

## Proposition 2 [Astolfi, Alessandri & Zaccarian, IEEE TAC 2021]

- Suppose  $\mathbf{v}$  is constant.
- Suppose  $(A - LC)$  is Hurwitz.

For any  $\theta \geq \lambda$  the disturbance-to-error *DC*-gains between  $\mathbf{v}$  and  $|\tilde{x}_0|$ , and between  $\mathbf{v}$  and  $|\tilde{x}_{\text{sat}}|$ , coincide.

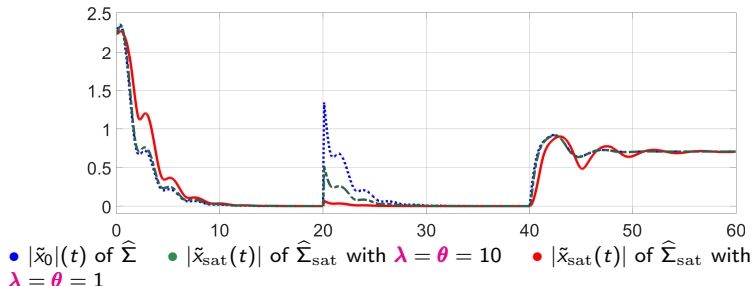
- For constant  $\mathbf{v}$ , the redesigned observer cannot do worse than the nominal one.

## A Numerical Example

$$\Sigma : \begin{cases} \dot{x} = Ax + Bu \\ y = Cx + v \end{cases} \quad \hat{\Sigma}_{\text{sat}} : \begin{cases} \dot{\hat{x}} = A\hat{x} + Bu + L \text{ sat}_{\sigma}(y - C\hat{x}) \\ \dot{\sigma} = -\lambda \sigma + \theta |y - C\hat{x}| \end{cases}$$

$$A = \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}, \quad B = \begin{pmatrix} 0 \\ 0 \end{pmatrix}, \quad C = (1 \quad 0), \quad L = \begin{pmatrix} 1 \\ 1 \end{pmatrix}, \quad x(0) = \begin{pmatrix} 0 \\ 1 \end{pmatrix}, \quad \hat{x}(0) = \begin{pmatrix} 1 \\ -1 \end{pmatrix}$$

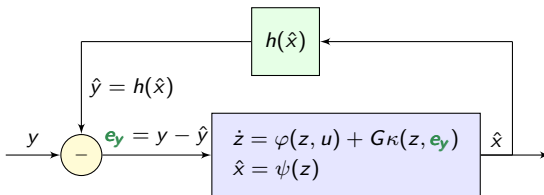
$$v(t) = \begin{cases} 0 & 0 \leq t \leq 20, \\ \delta_{\tau}(t) & 20 \leq t \leq 40, \\ 1 & 40 \leq t \leq 60, \end{cases} \quad \tau = 0.01$$



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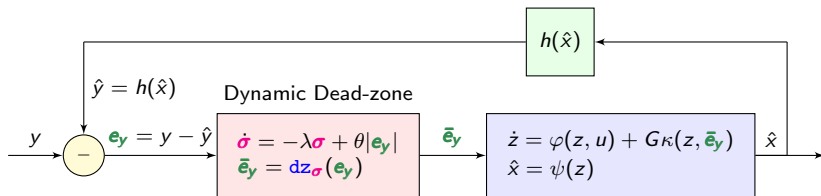
## Another scenario: persistent small measurement noise



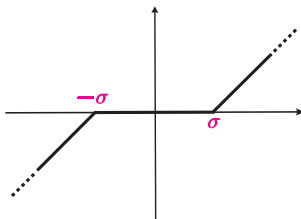
- Suppose  $y = h(x) + \mathbf{v}$ , with  $\mathbf{v}$  a small persistent measurement noise.
- When  $\mathbf{e}_y$  is persistently small, we have  $\mathbf{e}_y \simeq \mathbf{v}$ :
  - $\Rightarrow$  the information given by  $\mathbf{e}_y$  is not reliable
  - $\Rightarrow$  we want to trim out  $\mathbf{e}_y$ .

# A Dynamic Dead-Zone Scheme

We modify the previous structure by adding a dynamic dead-zone for  $\mathbf{e}_y = y - \hat{y}$ , i.e.,



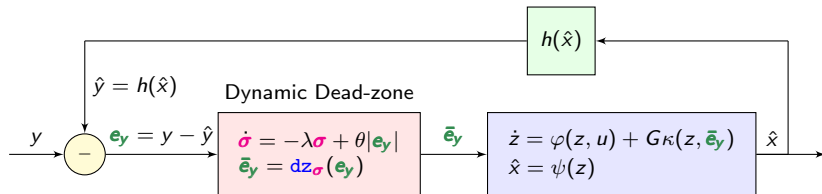
where  $\mathbf{dz}_{\sigma}(s) := s - \text{sat}_{\sigma}s$  and  $\lambda, \theta > 0$ .





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where  $\mathbf{dz}_\sigma(s) := s - \mathbf{sat}_\sigma s$  and  $\lambda, \theta > 0$ .

- The level of the dead-zone is selected as the current value of  $\sigma$ .
- If  $|\mathbf{e}_y| > \sigma$  then  $|\mathbf{dz}_\sigma(\mathbf{e}_y)| = |\mathbf{e}_y| \left(1 - \frac{\sigma}{|\mathbf{e}_y|}\right) \leq |\mathbf{e}_y|$ .
- If  $|\mathbf{e}_y| < \sigma$  then  $\mathbf{dz}_\sigma(\mathbf{e}_y) = 0$ .
- We obtain a trimming effect of  $\mathbf{e}_y$ .

# Dead-Zone Redesign: Main Result

## Theorem 2 [Astolfi, Alessandri & Zaccarian, IEEE TAC 2021]

Consider a given system  $\Sigma$  and observer  $\hat{\Sigma}$

$$\Sigma : \begin{cases} \dot{x} = f(x, u) + w \\ y = h(x) + v \end{cases} \quad \hat{\Sigma} : \begin{cases} \dot{z} = \varphi(z, u) + G\kappa(z, y - h(\hat{x})) + Gd \\ \hat{x} = \psi(z). \end{cases}$$

Suppose that  $\hat{\Sigma}$  is an ISS observer for  $\Sigma$  satisfying the output-growth condition. Then, for any  $\theta > 0$  there exists a  $\lambda^* > 0$  such that, for any  $\lambda > \lambda^*$ , the observer

$$\hat{\Sigma}_{dz} : \begin{cases} \dot{z} = \varphi(z, u) + G\kappa(z, dz_{\sigma}(y - h(\hat{x}))) + Gd \\ \dot{\sigma} = -\lambda\sigma + \theta|y - h(\hat{x})| \\ \hat{x} = \psi(z) \end{cases}$$

is an ISS observer for  $\Sigma$  with ISS Lyapunov function satisfying the output-growth condition.

- If  $y \in \mathbb{R}^m$ ,  $m \geq 1$ , then  $\hat{\Sigma}_{dz}$  reads

$$\begin{aligned} \dot{z} &= \varphi(z, u) + G\kappa\left(z, [dz_{\sigma_1}(y_1 - h_1(\hat{x})), \dots, dz_{\sigma_m}(y_m - h_m(\hat{x}))]\right) \\ \dot{\sigma}_i &= -\lambda_i\sigma_i + \theta_i|y_i - h_i(\hat{x})| \quad i = 1, \dots, m. \end{aligned}$$

- In general,  $\lambda \geq \theta$ .

## Sketch of Proof of Theorem 2

- Rewrite the observer  $\widehat{\Sigma}_{dz}$  as

$$\begin{cases} \dot{\sigma} = -\lambda\sigma + \theta|y - h(\hat{x})| \\ \dot{z} = \varphi(z, u) + G\kappa(z, y - h(\hat{x})) + G\delta \\ \delta = \kappa(z, dz_{\sigma}(y - h(\hat{x}))) - \kappa(z, y - h(\hat{x})). \end{cases}$$

- In light of the Lipschitz properties of  $\kappa$  and the dead-zone function, we have

$$|\delta| \leq \bar{\kappa} |dz_{\sigma}(y - h(\hat{x})) - (y - h(\hat{x}))| \leq \bar{\kappa} \sigma.$$

- Use the Lyapunov function

$$(x, z, \sigma) \mapsto W(x, z, \sigma) = V(x, z) + \zeta \sigma$$

with  $\zeta > 0$ .

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$$(x, z, \sigma) \mapsto W(x, z, \sigma) = V(x, z) + \zeta\sigma$$

with  $\zeta > 0$ .

- When  $|y - h(\hat{x})| > \sigma$  then  $\delta = 0$  and we have

$$D^+W \leq -cV(x, z) - \zeta\lambda\sigma + \zeta\theta|y - h(\hat{x})|$$

using the output-growth condition

$$D^+W \leq -(c - \zeta\theta\ell_0)V(x, z) - \zeta\lambda\sigma \leq 0$$

for  $\zeta$  small enough ( $\theta$  is fixed).

## Sketch of Proof of Theorem 2

- Rewrite the observer  $\hat{\Sigma}_{dz}$  as

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- Use the Lyapunov function

$$(x, z, \sigma) \mapsto W(x, z, \sigma) = V(x, z) + \zeta\sigma$$

with  $\zeta > 0$ .

- When  $|y - h(\hat{x})| \leq \sigma$  then  $\delta \neq 0$  and we have

$$D^+W \leq -cV(x, z) + c_d|\delta| - \zeta\lambda\sigma + \zeta\theta|y - h(\hat{x})|$$

using the output-growth condition

$$D^+W \leq -cV(x, z) - (\zeta\lambda - \zeta\theta - \bar{\kappa}c_d)\sigma \leq 0$$

for  $\lambda$  large enough ( $\zeta, \theta$  are fixed).

## Sketch of Proof of Theorem 2

- Rewrite the observer  $\widehat{\Sigma}_{dz}$  as

$$\begin{cases} \dot{\sigma} = -\lambda \sigma + \theta |y - h(\hat{x})| \\ \dot{z} = \varphi(z, u) + G\kappa(z, y - h(\hat{x})) + G\delta \\ \delta = \kappa(z, dz_{\sigma}(y - h(\hat{x}))) - \kappa(z, y - h(\hat{x})). \end{cases}$$

- In light of the Lipschitz properties of  $\kappa$  and the dead-zone function, we have

$$|\delta| \leq \bar{\kappa} |dz_{\sigma}(y - h(\hat{x})) - (y - h(\hat{x}))| \leq \bar{\kappa} \sigma.$$

- Use the Lyapunov function

$$(x, z, \sigma) \mapsto W(x, z, \sigma) = V(x, z) + \zeta \sigma$$

with  $\zeta > 0$ .

- ✓ We conclude that  $D^+W \leq -\varepsilon W$  for some  $\varepsilon > 0$ .
- ✓ The analysis can be done with  $w, v, d$  to show the desired ISS properties.
- ✓ It is not too hard to verify also the other properties for the redesigned observer.

# Noise effect analysis for linear systems

- Consider the linear case

$$\Sigma : \begin{cases} \dot{x} = Ax + Bu \\ y = Cx + v \in \mathbb{R} \end{cases} \quad \widehat{\Sigma} : \dot{\hat{x}} = A\hat{x} + Bu + \mathbf{L}(y - C\hat{x})$$

and the (nonlinear) redesigned observer

$$\widehat{\Sigma}_{\text{dz}} : \begin{cases} \dot{\hat{x}} = A\hat{x} + Bu + \mathbf{L} \, \text{dz}_{\sigma}(y - C\hat{x}) \\ \dot{\sigma} = -\lambda\sigma + \theta|y - C\hat{x}|. \end{cases}$$

- Consider the error variables

$$\begin{aligned} \tilde{x}_0 &:= x - \hat{x}, & \hat{x} &\in \widehat{\Sigma}, \\ \tilde{x}_{\text{dz}} &:= x - \hat{x}, & \hat{x} &\in \widehat{\Sigma}_{\text{dz}}. \end{aligned}$$

- We are interested in analyzing the effect of constant perturbations (approximation of persistently small noise).

## Effect of constant noise in the linear case

### Proposition 3 [Astolfi, Alessandri & Zaccarian, IEEE TAC 2021]

- Suppose  $\mathbf{v}$  is constant.
- Suppose  $(A - \mathbf{L}C)$  is Hurwitz and  $CA^{-1}\mathbf{L} < 1$ .

The disturbance-to-error *DC*-gains between  $\mathbf{v}$  and  $|\tilde{x}_0|$ , denoted as  $k_0$ , and between  $\mathbf{v}$  and  $|\tilde{x}_{dz}|$ , denoted as  $k_{dz}$ , satisfy

$$k_{dz} \leq \left[ 1 - \tilde{k} \left( \frac{\theta}{\lambda} \right) \right] k_0$$

for any  $\lambda > \theta \geq 0$  and for some  $\tilde{k} \in \mathcal{K}$ .

- The DC-gain is always reduced thus improving the rejection to measurement noise.
- The condition  $CA^{-1}\mathbf{L} < 1$  is always verified if both  $A$  and  $(A - \mathbf{L}C)$  are Hurwitz.

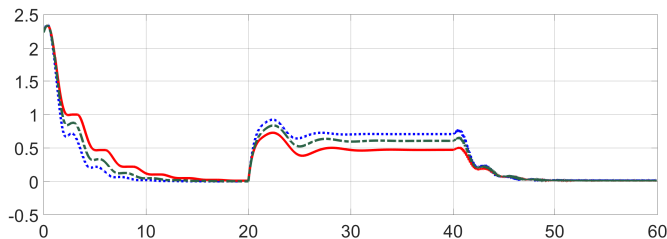


## A Numerical Example

$$\Sigma : \begin{cases} \dot{x} = Ax + Bu \\ y = Cx + v \end{cases} \quad \hat{\Sigma}_{\text{sat}} : \begin{cases} \dot{\hat{x}} = A\hat{x} + Bu + L dz_{\sigma}(y - C\hat{x}) \\ \dot{\sigma} = -\lambda \sigma + \theta |y - C\hat{x}| \end{cases}$$

$$A = \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}, \quad B = \begin{pmatrix} 0 \\ 0 \end{pmatrix}, \quad C = (1 \quad 0), \quad L = \begin{pmatrix} 1 \\ 1 \end{pmatrix}, \quad x(0) = \begin{pmatrix} 0 \\ 1 \end{pmatrix}, \quad \hat{x}(0) = \begin{pmatrix} 1 \\ -1 \end{pmatrix}$$

$$v(t) = \begin{cases} 0 & 0 \leq t \leq 20, \\ 1 & 20 \leq t \leq 40 \\ \sin(50t) & 40 \leq t \leq 60, \end{cases} \quad \begin{aligned} &\bullet \sup_{\infty} |\tilde{x}_0(t)| \leq 0.028 \\ &\bullet \sup_{\infty} |\tilde{x}_{dz}(t)| \leq 0.022 \\ &\bullet \sup_{\infty} |\tilde{x}_{dz}(t)| \leq 0.016 \end{aligned}$$



- $|\tilde{x}_0(t)|$  of  $\hat{\Sigma}$
- $|\tilde{x}_{dz}(t)|$  of  $\hat{\Sigma}_{\text{sat}}$  with  $\lambda = 4, \theta = 1$
- $|\tilde{x}_{dz}(t)|$  of  $\hat{\Sigma}_{\text{sat}}$  with  $\lambda = 2, \theta = 1$

## Application to vehicle lateral speed estimation

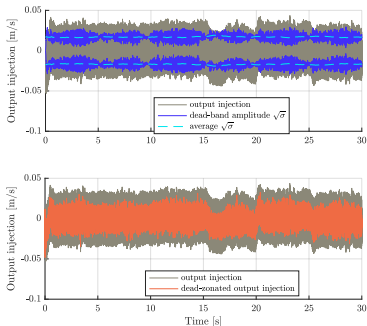
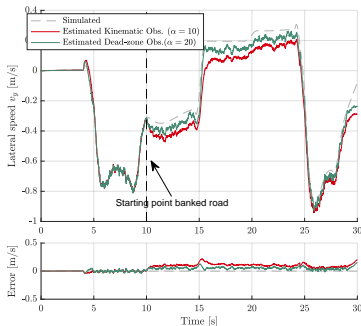
- linear parameter-varying model (parameter  $r$  = yaw rate)

$$\begin{cases} \dot{x} = A(r)x + u \\ y = Cx, \end{cases} \quad A(r) := \begin{bmatrix} 0 & r \\ -r & 0 \end{bmatrix}, \quad C := \begin{bmatrix} 1 & 0 \end{bmatrix}.$$

- yaw-rate dependent kinematic observer

$$\begin{cases} \dot{\hat{x}} = A(r)\hat{x} + u + L(r)(\hat{y} - y) \\ \hat{y} = C\hat{x}, \end{cases} \quad L(r) := \begin{bmatrix} -2\alpha|r| \\ (1 - \alpha^2)r \end{bmatrix},$$

- deadzonated observer reduces the sensitivity to noise

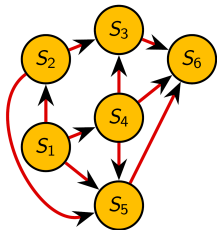


# Outline

- 1 Introduction
- 2 Observer Class
- 3 Stubborn Redesign
- 4 Dead-Zone Redesign
- 5 Synchronization**
- 6 Dynamic Output Feedback
- 7 Conclusions

# A synchronization problem for multi-agent systems

- Consider a synchronization problem among  $N$  identical agents



$$\dot{x}_i = Ax_i + \phi(x_i) + u_i \quad i = 1, \dots, N,$$

$$y_i = Cx_i \quad i = 1, \dots, N,$$

$$x_i \in \mathbb{R}^n, \quad u_i \in \mathbb{R}, \quad y_i \in \mathbb{R}.$$

- We want to achieve consensus among all the states

$$\lim_{t \rightarrow \infty} |x_i(t) - x_j(t)| = 0 \quad \forall i, j \in \{1, \dots, N\}.$$

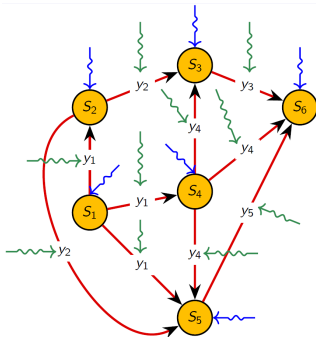
- We want to use a distributed control law.
- A typical solution is to use a diffusive coupling

$$u_i = -\mathbf{K} \sum_{j=1}^N \ell_{ij} y_j \quad \ell_{ij} \in L$$

where  $L$  is the Laplacian matrix of the graph.

# Perturbations in networks

- What happens in the presence of perturbations?



$$\dot{x}_i = Ax_i + \phi(x_i) + u_i$$

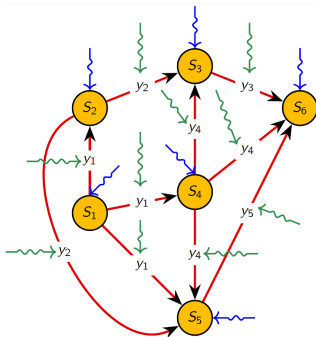
$$y_i = Cx_i + v_i$$

- If we have some ISS property then

$$\limsup_{t \rightarrow \infty} |x_i(t) - x_j(t)| \leq \gamma \left( \sum_{i=1}^N |v_i| \right) \quad \forall i, j \in \{1, \dots, N\} \quad \gamma \in \mathcal{K}.$$

# Perturbations in networks

- What happens in the presence of perturbations?



$$\dot{x}_i = Ax_i + \phi(x_i) + u_i$$

$$y_i = Cx_i + v_i$$

- Design of diffusing coupling for nonlinear systems in the presence of output perturbations  $v_i$ ?

# Saturation and Dead-Zone Redesign for Diffusing Coupling

- In the presence of impulsive disturbances (e.g. outliers, switching topologies, ...) we use a **dynamic saturation redesign**

$$\begin{aligned}\dot{\sigma}_i &= -\lambda \sigma_i + \theta_i \left| \sum_{j=1}^N \ell_{ij} y_j \right| & i = 1, \dots, N \\ u_i &= -K \text{sat}_{\sigma_i} \left( \sum_{j=1}^N \ell_{ij} y_j \right) & i = 1, \dots, N.\end{aligned}$$

- In the presence of persistent disturbances (communication networks noise) we use a **dynamic dead-zone redesign**

$$\begin{aligned}\dot{\sigma}_i &= -\lambda \sigma_i + \theta_i \left| \sum_{j=1}^N \ell_{ij} y_j \right| & i = 1, \dots, N \\ u_i &= -K \text{dz}_{\sigma_i} \left( \sum_{j=1}^N \ell_{ij} y_j \right) & i = 1, \dots, N.\end{aligned}$$

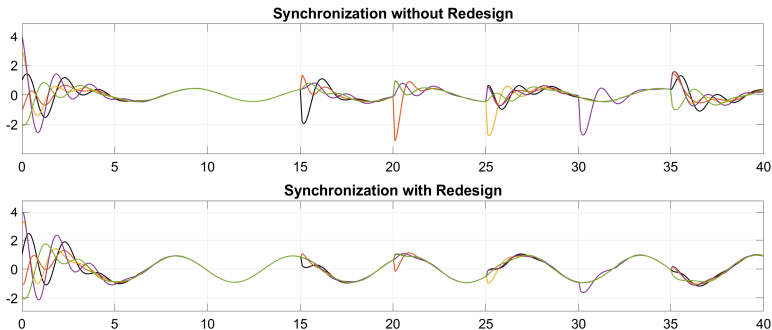
- Synchronization in nominal conditions  $\mathbf{v}_i = 0$  is preserved.
- Performance in the presence of noise is improved.
- Proofs and philosophy design are very similar to the observer design.

# Simulation Example for Saturation Redesign

Consider a network of  $N = 6$  linear oscillators

$$\dot{x}_i = \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} x_i + u_i, \quad y_i = (1 \quad 0) x_i + \mathbf{v}_i$$

in the presence of outliers  $\mathbf{v}_i$ .



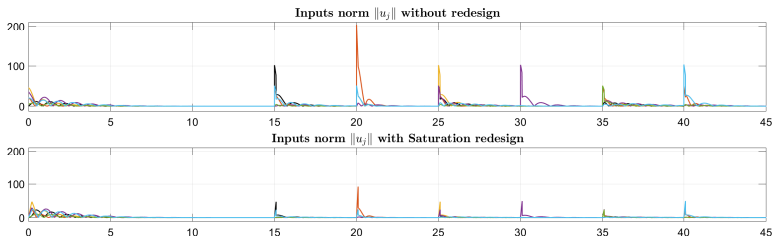


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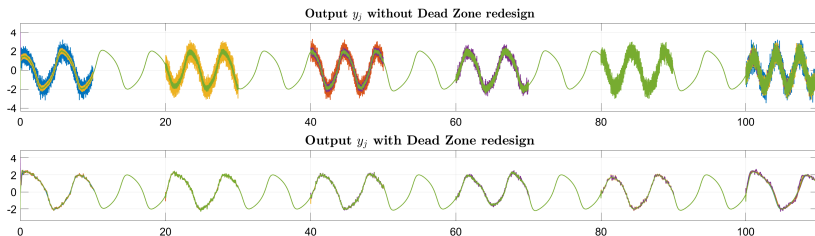


# Simulation Example for Dead-Zone Redesign

Consider a network of  $N = 5$  Van der Pol oscillators

$$\dot{x}_i = \begin{pmatrix} 0 & 1 \\ -1 & \mu \end{pmatrix} x_i + \begin{pmatrix} 0 \\ -\mu x_{i1}^2 x_{i2} \end{pmatrix} + u_i, \quad y_i = \begin{pmatrix} 1 & 0 \end{pmatrix} x_i + v_i$$

in the presence of white noise  $v_i$ .

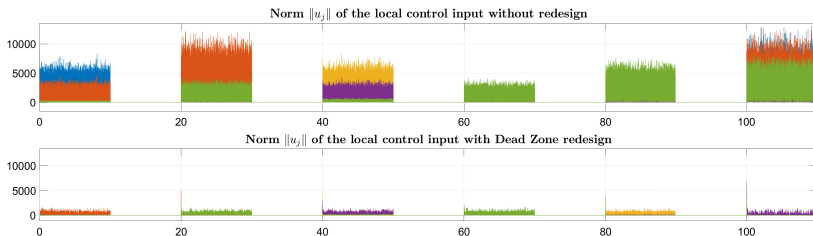


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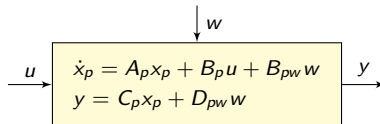
in the presence of white noise  $\mathbf{v}_i$ .



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# Dynamic Output Feedback

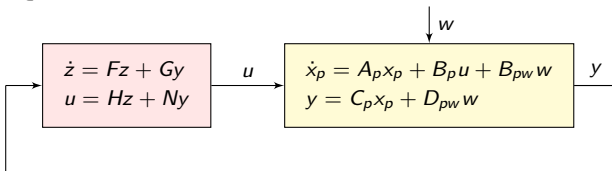


## Plant

- state  $x_p \in \mathbb{R}^{n_p}$
- known external input  $u \in \mathbb{R}^m$
- measured output  $y \in \mathbb{R}^p$
- unknown system and measurement disturbances  $w \in \mathbb{R}^{n_d}$

## Controller

- state  $z \in \mathbb{R}^{n_c}$



## Assumption 1

The linear closed-loop system with  $w \equiv 0$  is globally exponentially stable to the origin.

## Stubborn Redesign to handle Measurement Outliers

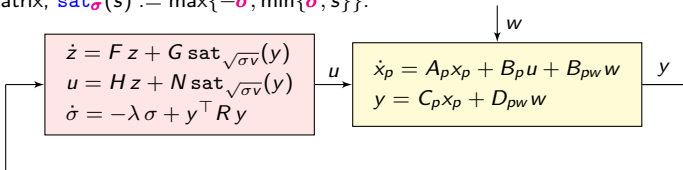
- The controller dynamics is augmented with a new non-negative state  $\sigma \in \mathbb{R}_{\geq 0}$ :

$$\dot{z} = Fz + G \text{sat}_{\sqrt{\sigma}v}(y)$$

$$u = Hz + N \text{sat}_{\sqrt{\sigma}v}(y)$$

$$\dot{\sigma} = -\lambda \sigma + y^\top R y$$

where  $\sqrt{\sigma}v$  are the componentwise square-roots of the elements of the non-negative vector  $v$  scaled by  $\sigma$ ;  $\lambda > 0$ ;  $R$  is a symmetric, positive definite matrix;  $\text{sat}_\sigma(s) := \max\{-\sigma, \min\{\sigma, s\}\}$ .



- Notice:
  - ✓ The level of the saturation is a scaled square root of the current value of  $\sigma$ .
  - ✓ The value of  $\sigma$  is dynamically adapted according to  $y^\top R y$ .
- What happens in presence of an outlier?
  - ✓ If  $y$  is persistently constant, then  $\sigma$  tends to a constant value.
  - ✓ If an outlier occurs,  $y$  becomes large but the effect of  $y$  on the loop is saturated on the current (small) value of  $\sqrt{\sigma}v$ , mitigating its effect on the closed-loop system.
  - ✓ If  $y$  grows or is persistently affected by outliers, then  $\sigma$  increases, thus generating desaturation.

## Closed-Loop Input-to-State Stability After Redesign

Theorem 3 [Tarbouriech, Alessandri, Astolfi, Zaccarian, IEEE LCSS 2022]

Assume that there exist a scalar  $\lambda > 0$ , a symmetric positive definite matrix  $P \in \mathbb{R}^{n \times n}$ , a symmetric positive semi-definite matrix  $R \in \mathbb{R}^{p \times p}$ , two diagonal positive definite matrices  $U_g \in \mathbb{R}^{p \times p}$ ,  $U_\ell \in \mathbb{R}^{p \times p}$ , and a matrix  $Y \in \mathbb{R}^{p \times n}$  such that inequalities

$$M_g := \text{He} \begin{bmatrix} PA - \frac{1}{2}\lambda C^\top RC + \frac{1}{2}\lambda P & -PB \\ U_g C & -U_g \end{bmatrix} < 0$$

$$M_\ell := \text{He} \begin{bmatrix} PA & -PB \\ U_\ell C + Y & -U_\ell \end{bmatrix} < 0$$

$$\begin{bmatrix} P & Y_{(i)}^\top \\ Y_{(i)} & \lambda^{-1} u_{\ell,i} \end{bmatrix} \geq 0, \quad \forall i = 1, \dots, p,$$

are satisfied, where  $\text{He}(\star) := \star + \star^\top$ . Then, the closed loop system with dynamically **saturated output regulator** having entries with a vector  $v$  as the inverse of the diagonal elements of  $U_\ell$  (namely  $\text{diag}(v) = U_\ell^{-1}$ ), is finite-gain exponentially ISS from  $w$  to  $x$ , namely there exist positive scalars  $M$ ,  $\alpha > 0$  and  $\gamma$  such that all solutions satisfy

$$\left| (x(t), \sqrt{\sigma(t)}) \right| \leq M e^{-\alpha t} \left| (x(0), \sqrt{\sigma(0)}) \right| + \gamma \sup_{\tau \in [0, t]} |w(\tau)|,$$

for all  $t \geq 0$ .

# Sketch of Proof and Feasibility

- The proof of Theorem 3 follows from using the Lyapunov function

$$\mathcal{V}(x, \sigma) = x^\top P x + \zeta \sigma + \mu \max\{x^\top P x - \lambda \sigma, 0\}$$

with  $P$  symmetric positive semi-definite matrix and constants  $\zeta, \mu > 0$  to be suitably chosen.

- It is an ISS Lyapunov function and it is **not continuously differentiable**.
- For an overview on nonsmooth Lyapunov functions, see [Della Rossa, Goebel, Tanwani, Zaccarian, "Piecewise structure Lyapunov functions and densely checked decrease conditions for hybrid systems," MCSS 2021].
- The LMIs involved by Theorem 3 are homogeneous, thus they can be solved with the additional condition  $P > I$  for increased numerical robustness

**Proposition 4** [Tarbouriech, Alessandri, Astolfi, Zaccarian, IEEE LCSS 2022]

Under Assumption 1 there exist parameters  $P, R, U_\ell, U_g, Y$  and  $\lambda$  satisfying the conditions of Theorem 3.



## Simulation Case Study

- Linearization about an equilibrium point of the **longitudinal dynamics of fixed-wing aircrafts** flying at high speed [Astolfi, Praly, "Integral action in output feedback for multi-input multi-output nonlinear systems," IEEE TAC, 2017]:

$$\begin{aligned}\dot{v} &= e - g \sin(\gamma) + w_1 \\ \dot{\gamma} &= \ell v \sin(\theta - \gamma) - \frac{g \cos(\gamma)}{v} \\ \dot{\theta} &= q\end{aligned}$$

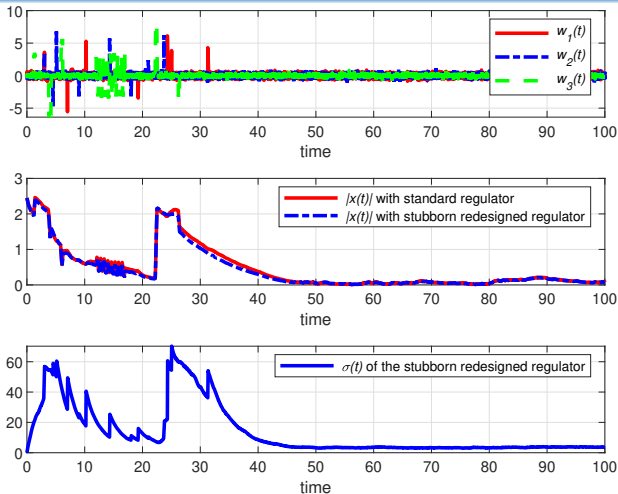
where  $v$  is the modulus of the speed,  $\gamma$  is the flight path angle,  $\theta$  is the pitch angle,  $q$  is the pitch rate,  $e$  is the propulsive balance,  $g$  is the standard gravitational acceleration,  $\ell$  is an aerodynamic lift coefficient,  $w_1$  is a wind perturbation.

- The signals  $e, q$  are regarded as control inputs and  $\gamma, \theta$  as measured outputs.
- The measurement noises  $w_2, w_3$  affect the outputs.
- The linearization around an equilibrium  $(v_0, 0, 0)$  of this model provides matrices  $A_p, B_p, C_p$  as follows:

$$\left[ \begin{array}{c|c|c} A_p & B_p & B_{pw} \\ \hline C_p & & D_{pw} \end{array} \right] = \left[ \begin{array}{ccc|cc|ccc} 0 & -g & 0 & 1 & 0 & 1 & 0 & 0 \\ g v_0^{-2} & -\ell v_0 & \ell v_0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ \hline 0 & 1 & 0 & & & 0 & 1 & 0 \\ 0 & 0 & 1 & & & 0 & 0 & 1 \end{array} \right].$$

- We choose  $g = 1$ ,  $v_0 = 2$ ,  $\ell = 0.1$  and used pole placement to select closed-loop poles having real part in  $[-3, -0.1]$ .

# Simulation Results



Integrals of $ x(t) $ , $t \in [0, 60]$	
Standard reg.	Stubborn reg.
42.4572	39.4415

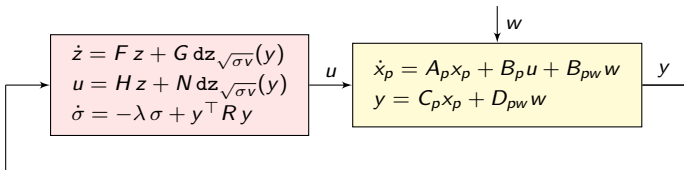
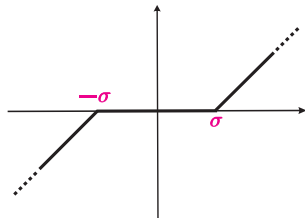
- ✓ The adaptive saturation attenuates the effect of the outliers upon their sudden rise.
- ✗ Small noises are not reduced at steady state.

# Deadzone Redesign to handle persistent measurement noise

- The controller dynamics, augmented with a new non-negative state  $\sigma \in \mathbb{R}_{\geq 0}$ , is given by

$$\begin{aligned}
 \dot{z} &= Fz + G \, \text{dz}_{\sqrt{\sigma v}}(y) \\
 u &= Hz + N \, \text{dz}_{\sqrt{\sigma v}}(y) \\
 \dot{\sigma} &= -\lambda \sigma + y^\top R y,
 \end{aligned}$$

where we modify the original structure by adding a dynamic dead-zone.



- ✓ The dead-zone provides a trimming effect on  $y$ , which denoises the feedback loop when the output is close enough to zero.

# Closed-Loop Input-to-State Stability After Redesign

Theorem 5 [Tarbouriech, Alessandri, Astolfi, Zaccarian, IEEE LCSS 2022]

Assume that there exist a scalar  $\lambda > 0$ , a symmetric positive definite matrix  $P \in \mathbb{R}^{n \times n}$ , a symmetric positive semi-definite matrix  $R \in \mathbb{R}^{p \times p}$ , a diagonal positive definite matrices  $U_g \in \mathbb{R}^{p \times p}$  such that

$$M_g := \text{He} \begin{bmatrix} PA + \frac{1}{2}C^\top RC & -PB \\ U_g C & -U_g(1 + \lambda) \end{bmatrix} < 0,$$

is satisfied, where  $\text{He}(\star) := \star + \star^\top$ . Then, the closed loop system with dynamically **dead-zonated output regulator** having entries with a vector  $v$  as the inverse of the diagonal elements of  $U_g$  (namely  $\text{diag}(v) = U_g^{-1}$ ), is finite-gain exponentially ISS from  $w$  to  $x$ , namely there exist positive scalars  $M, \alpha > 0$  and  $\gamma$  such that all solutions satisfy

$$\left| (x(t), \sqrt{\sigma(t)}) \right| \leq M e^{-\alpha t} \left| (x(0), \sqrt{\sigma(0)}) \right| + \gamma \sup_{\tau \in [0, t]} |w(\tau)|,$$

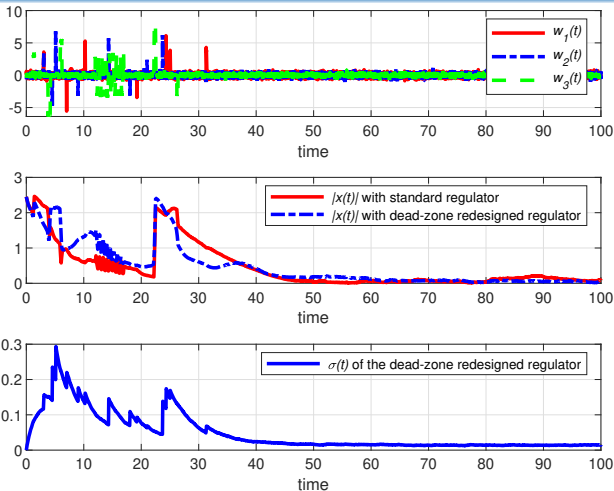
for all  $t \geq 0$ .

- The proof of Theorem 5 follows from using the Lyapunov function  $\mathcal{V}(x, \sigma) = x^\top P x + 2\sigma$  with  $P = P^\top > 0$  to be suitably chosen.

Proposition 6 [Tarbouriech, Alessandri, Astolfi, Zaccarian, IEEE LCSS 2022]

Under Assumption 1 there exist parameters  $P, R, U_g$ , and  $\lambda$  satisfying the conditions of Theorem 5.

# Simulation Results



Integrals of $ x(t) $ , $t \in [60, 100]$	
Standard reg.	Dead-zone reg.
3.6459	2.5558

- ✓ The dead-zone adaptation attenuates the effect of small noises at steady state.
- ✗ Outliers deteriorate performance.

# Outline

- 1 Introduction
- 2 Observer Class
- 3 Stubborn Redesign
- 4 Dead-Zone Redesign
- 5 Synchronization
- 6 Dynamic Output Feedback
- 7 Conclusions**

# Conclusions

## Summary

- Adaptive nonlinearities such as **saturation and dead-zone** can **improve the performance** of state observers and controllers in the presence of measurement noise.
- General and flexible approach to **redesign** ISS observers and output feedback regulators while **preserving ISS**.
- The effectiveness of the redesign approach follows from devising the **appropriate ISS Lyapunov functions**.
- For linear systems, **design conditions based on LMIs** are established that can be treated by means of convex optimization tools.

## Next Goals

- redesign in output feedback control for **nonlinear continuous-time systems**;
- extension of the redesign approach for estimation and control of **discrete-time systems**;
- experimental validation of the approach on suitable **case studies**.