Linear flux observers for induction motors with quadratic Lyapunov certificates

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Main Equations of the Induction Motor LTV model

State space model of the induction motor is:

$$\dot{x}(t) = A(t)x(t) + Bu(t),
y(t) = Cx(t),$$

where x, u and y are respectively the state, the input and the output vectors defined as:

$$\mathbf{x} = \begin{bmatrix} i_{\mathsf{sd}} & i_{\mathsf{sq}} & \psi_{\mathsf{rd}} & \psi_{\mathsf{rq}} \end{bmatrix}^\mathsf{T}; \ \mathbf{y} = \begin{bmatrix} i_{\mathsf{sd}} & i_{\mathsf{sq}} \end{bmatrix}^\mathsf{T}; \ \mathbf{u} = \begin{bmatrix} u_{\mathsf{sd}} & u_{\mathsf{sq}} \end{bmatrix}^\mathsf{T},$$

denoting direct and quadrature stator voltage u_{s*} and current i_{s*} and rotor flux $\psi_{r\star}$, and

$$A(t) = \begin{bmatrix} -\gamma & 0 & \alpha\beta & \beta\omega_{re}(t) \\ 0 & -\gamma & -\beta\omega_{re}(t) & \alpha\beta \\ \alpha L_m & 0 & -\alpha & -\omega_{re}(t) \\ 0 & \alpha L_m & \omega_{re}(t) & -\alpha \end{bmatrix}; B = \begin{bmatrix} \frac{1}{\sigma L_s} & 0 \\ 0 & \frac{1}{\sigma L_s} \\ 0 & 0 \\ 0 & 0 \end{bmatrix}; C = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}^T.$$

Parameters σ , α , β and γ , are defined as:

$$\sigma = 1 - \frac{L_m^2}{L_s L_r} \in (0, 1), \quad \alpha = \frac{R_r}{L_r} > 0, \quad \beta = \frac{L_m}{\sigma L_s L_r} > 0, \quad \gamma = \frac{R_s}{\sigma L_s} + \beta \alpha L_m > 0. \tag{PARS}$$

With matrices

Full-order Design

$$A(t) = \begin{bmatrix} -\gamma & 0 & \alpha\beta & \beta\omega_{re}(t) \\ 0 & -\gamma & -\beta\omega_{re}(t) & \alpha\beta \\ \alpha L_m & 0 & -\alpha & -\omega_{re}(t) \\ 0 & \alpha L_m & \omega_{re}(t) & -\alpha \end{bmatrix}; B = \begin{bmatrix} \frac{1}{\sigma L_s} & 0 \\ 0 & \frac{1}{\sigma L_s} \\ 0 & 0 \\ 0 & 0 \end{bmatrix}; C = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}^T,$$

the linear time-varying (LTV) equations (\spadesuit), can be written as (with J =

$$\dot{x}(t) = \left(\overbrace{\begin{bmatrix} -\gamma & \alpha\beta \\ \alpha L_m & -\alpha \end{bmatrix}}^{\bar{A}} \otimes I + \overbrace{\begin{bmatrix} 0 & -\beta\omega_{re}(t) \\ 0 & \omega_{re}(t) \end{bmatrix}}^{\Omega(t)} \otimes J \right) x(t) + \left(\overbrace{\begin{bmatrix} \frac{1}{\sigma L_s} \\ 0 \end{bmatrix}}^{\bar{B}} \otimes I \right) u(t),$$

$$y(t) = \left(\overbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}^{\bar{C}} \otimes I \right) x(t),$$

The simplified LTI model naturally arising from original LTV model is:

$$\dot{\bar{x}}(t) = \bar{A}\bar{x}(t) + \bar{B}u(t), \quad \bar{y}(t) = \bar{C}\bar{x}(t).$$
 (Areduced)

Lemma

Matrix \bar{A} is Hurwitz and triple $(\bar{A}, \bar{B}, \bar{C})$ is controllable and observable for any value of the physical parameters satisfying (PARS).

Let us consider the following LTV observer dynamics:

$$\dot{\hat{x}} = \left[\bar{A} \otimes I + \Omega(t) \otimes J\right] \hat{x} + \left(\bar{B} \otimes I\right) u + \left[\bar{L} \otimes I + \begin{bmatrix} 0 \\ \rho \omega_{re}(t) \end{bmatrix} \otimes J\right] (y - \hat{y}).$$

Theorem (Main)

Consider any constant gain \bar{L} such that $\bar{A} - \bar{L}\bar{C}$ is Hurwitz and any pair of positive definite matrices $\bar{P} = \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix}$, \bar{Q} such that:

$$\operatorname{He}\left(\bar{P}(\bar{A}-\bar{L}\bar{C})\right) = \left(\bar{P}(\bar{A}-\bar{L}\bar{C})\right) + \left(\bar{P}(\bar{A}-\bar{L}\bar{C})\right)^{T} \le -\bar{Q} < 0. \tag{\$}$$

Then, selecting $\rho = \frac{\beta p_{11} - p_{12}}{p_{22}}$, and denoting the estimation error as $e = x - \hat{x}$, the following quadratic Lyapunov conditions hold:

$$\begin{split} V(e) &= \frac{1}{2} e^T \left(\bar{P} \otimes I \right) e = e^T \begin{bmatrix} \frac{p_{11} & 0 & p_{12} & 0}{0 & p_{11} & 0 & p_{12} \\ 0 & p_{11} & 0 & p_{12} & 0 \\ p_{12} & 0 & p_{22} & 0 \\ 0 & p_{12} & 0 & p_{22} \end{bmatrix} e \text{ is positive definite} \\ \dot{V}(e) &= \langle \nabla V(e), \dot{e} \rangle = -e^T \left(\bar{Q} \otimes I \right) e, \end{split}$$

along all solutions to (\spadesuit) , (\heartsuit) for any time-varying $t \mapsto \omega_{re}(t)$.

Interpretations of Theorem Main

We can design the gain \bar{L} for the 2×2 LTI model (\spadesuit reduced) and then we obtain the same features for the 4×4 LTV model (\spadesuit).

Clearly, the error variables e depend on ω_{re} and exhibit a peculiar time-varying transient, but the upper bound on V(e) is a simple exponential function.

Tight upper bound by solving the convex optimization with α_ℓ being the spectral abscissa of $\bar{A} - \bar{L}\bar{C}$ (namely $\alpha_\ell = -\max_i \left(\operatorname{Re}\{\lambda_i(\bar{A} - \bar{L}\bar{C})\} \right)$),

$$\min_{k,ar{P}=ar{P}^T}k, \;\; ext{subject to:} \ \operatorname{He}\left(ar{P}(ar{A}-ar{L}ar{C})
ight)\leq -2lpha_\ellar{P}, \qquad (=-ar{Q}) \ I\leq ar{P}\leq kI,$$

to obtain the lifted bound:

$$|e(t)| \leq \sqrt{k} e^{-\alpha_{\ell} t} |e(0)|, \quad \forall t \geq 0$$

This result follows from Theorem 2 applied with $\bar{Q}=2\alpha_{\ell}\bar{P}$.



Full-order Design

Observer gain \overline{L} selection for \mathcal{L}_2 gain minimization

Gain \bar{L} can be chosen to **minimize** the \mathcal{L}_2 gain between a disturbance d (acting on the current measurement) and the estimation error e.

Theorem **Main** can be applied with $\rho = 0$, which leads to the following error dynamics:

$$\dot{e} = ((\bar{A} - \bar{L}\bar{C}) \otimes I + \Omega(t) \otimes J) e + (\bar{L} \otimes I) d.$$
 (ERR)

An upper bound μ on the \mathcal{L}_2 gain from d to e for dynamics (ERR) can be minimized by solving the LMI formulation of the Bounded Real Lemma [1]¹

$$\begin{split} \min_{\mu,\bar{P},\bar{X}} \mu, & \text{ subject to:} \\ \bar{P} &= \begin{bmatrix} \frac{\rho_{11}}{\beta \rho_{11}} \frac{\beta \rho_{11}}{\rho_{22}} \end{bmatrix} > 0, \\ \operatorname{He} \begin{bmatrix} \bar{P}\bar{A}-\bar{X}\bar{C} & -\bar{X} & 0 \\ 0 & -\frac{\mu}{2}I & 0 \\ I & 0 & -\frac{\mu}{2}I \end{bmatrix} < 0, \\ \operatorname{He} \left(\bar{P}\bar{A}-\bar{X}\bar{C}\right) \leq -2\alpha_{\mathsf{des}}\bar{P}, \end{split}$$

and then selecting $\bar{L} = \bar{P}^{-1}\bar{X}$, where $\alpha_{\rm des} > 0$ is any desired convergence rate

¹[1] G. Dullerud, F. Paganini, ACourse in Robust Control Theory. Springer, 2000.



Explicit selections of observer gain \bar{L} and certificate \bar{P}

A few relevant explicit selections of \bar{L} can be given for Theorem Main:

1 (Open-loop observer) Selection:

$$\bar{L} = \begin{bmatrix} 0 & 0 \end{bmatrix}^T, \quad \bar{P} = \begin{bmatrix} \frac{\sigma L_s L_r}{0} \\ 0 & 1 \end{bmatrix} > 0, \quad \bar{Q} = 2 \begin{bmatrix} \frac{\gamma \sigma L_s L_r}{-\alpha L_m} \\ -\alpha L_m & \alpha \end{bmatrix} > 0, \text{ (Lzero)}$$

is such that $\bar{A} - \bar{L}\bar{C}$ is Hurwitz and \bar{P} , \bar{Q} satisfy (\clubsuit).

2 (Speed of convergence α) Selection:

$$\bar{L} = \begin{bmatrix} \alpha - \gamma & \alpha L_m \end{bmatrix}^T, \quad \bar{P} = \begin{bmatrix} \frac{1}{\alpha\beta} & 0 \\ 0 & \frac{\beta}{\alpha} \end{bmatrix} > 0, \quad \bar{Q} = \begin{bmatrix} \frac{2}{\beta} & -1 \\ -1 & 2\beta \end{bmatrix} > 0,$$

is such that $\bar{A} - \bar{L}\bar{C}$ is Hurwitz and assigns both eigenvalues of $\bar{A} - \bar{L}\bar{C}$ at $-\alpha$. Moreover selections \bar{P} , \bar{Q} satisfy (\clubsuit).

3 (Arbitrary speed of convergence $(\alpha + \eta)$). Given any scalar $\eta > 0$, selection:

$$\bar{L} = \begin{bmatrix} \alpha - \gamma + 2\eta \\ \alpha L_m + \frac{\eta}{\beta} \left(1 + 2\frac{\eta}{\alpha} \right) \end{bmatrix}, \quad \bar{P} = \begin{bmatrix} \frac{\eta}{\alpha} \left(1 + 2\frac{\eta}{\alpha} \right) & -\frac{\beta}{\alpha} \eta \\ -\frac{\beta}{\alpha} \eta & \beta^2 \end{bmatrix}, \quad \bar{Q} = 2(\alpha + \eta)\bar{P}$$
(Uspeed)

is such that $\bar{A} - \bar{L}\bar{C}$ is Hurwitz and \bar{P} , \bar{Q} satisfy (\clubsuit).



Reduced order observer

Given any gain \bar{L} and any matrices $\bar{P} = \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} > 0$, $\bar{Q} = \begin{bmatrix} q_{11} & q_{12} \\ q_{12} & q_{22} \end{bmatrix} > 0$ satisfying Theorem Main, we introduce the reduced order observer [2, Lemma 3.1]2:

$$\dot{\hat{\phi}} = A_{\psi}(t) \begin{bmatrix} y \\ \hat{\psi} \end{bmatrix} + \rho_{22}^{-1} \rho_{12} \left(A_{i}(t) \begin{bmatrix} y \\ \hat{\psi} \end{bmatrix} + \frac{1}{\sigma L_{s}} u \right)$$

$$\dot{\psi} = \hat{\phi} - \rho_{22}^{-1} \rho_{12} y, \qquad A(t) = \begin{bmatrix} \frac{A_{i}(t)}{A_{\psi}(t)} \end{bmatrix} = \begin{bmatrix} \frac{-\gamma I - \alpha \beta I - \beta \omega_{re}(t)J}{\alpha L_{m}I - \alpha I + \omega_{re}(t)J} \end{bmatrix}$$
(Ored)

Proposition (Reduced Order Observer)

If matrices (\bar{P}, \bar{Q}) , and gain \bar{L} satisfy (\clubsuit) , then the flux estimation error $e_{\psi} = \psi - \hat{\psi}$ satisfies the following quadratic Lyapunov conditions:

$$egin{aligned} V_{\psi}(e_{\psi}) &= rac{1}{2} e_{\psi}^{-T}(p_{22} \otimes I) e_{\psi} ext{ is positive definite} \ \dot{V}_{\psi}(e_{\psi}) &= -lpha \left(1 - eta
ho_{22}^{-1} p_{12}
ight) V_{\psi}(e_{\psi}), \end{aligned}$$

along dynamics (\spadesuit) , $(\heartsuit red)$.

²[2] G. Besancon, Remarks on nonlinear adaptive observer design. Systems & control letters, 41(4), pp 271-280, 2000.

Comparison with works presented in the literature

With selection (Lspeed), if $\eta = \beta \alpha c \implies p_{22}^{-1} p_{12} = -c$, \implies the reduced observer coincides with [3, Equation (3.33)]³:

$$egin{aligned} \dot{\hat{\phi}} &= \left(\left[-lpha (1+eta c) & lpha L_m + c(\gamma - lpha - lpha eta)
ight] \otimes I + \\ &+ \left[\omega_{re}(t) (1+eta c) & c\omega_{re}(t) (1+eta c)
ight] \otimes J
ight) \left[egin{aligned} \hat{\phi} \ i \end{aligned}
ight] - rac{c}{\sigma L_s} u, \ \hat{\psi} &= \hat{\phi} + ci. \end{aligned}$$

With selection (Lzero), $\Longrightarrow p_{22}^{-1}p_{12} = 0$,

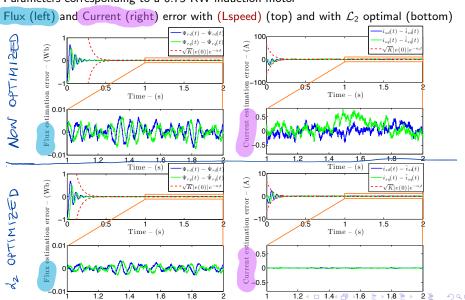
 \implies the reduced-order observer coincides with observer [3, equation (3.8)]³:

$$\dot{\hat{\phi}} = \left(\begin{bmatrix} -\alpha & \alpha L_m \end{bmatrix} \otimes I + \begin{bmatrix} \omega_{re}(t) & 0 \end{bmatrix} \otimes J \right) \begin{bmatrix} \hat{\phi} \\ i \end{bmatrix},$$
 $\hat{\psi} = \hat{\phi}.$

³[3] Riccardo Marino, Patrizio Tomei, and Cristiano M Verrelli. *Induction motor control design.* Springer, 2010.

Simulations show improved results with \mathcal{L}_2 optimized gains

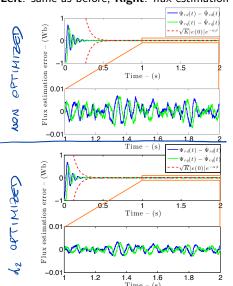
Parameters corresponding to a 0.75 KW induction motor

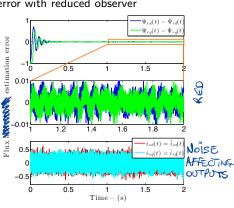


Reduced-order solution: deteriorated flux estimation e,

Parameters corresponding to a 0.75 KW induction motor

Left: same as before, Right: flux estimation error with reduced observer

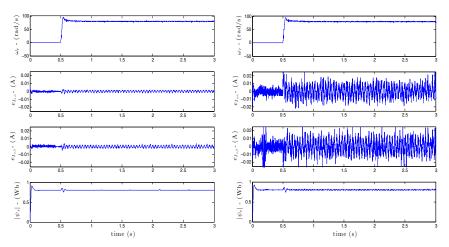






Experimental results confirm advantage of \mathcal{L}_2 optimization

Field Oriented Control law on an experimental 0.75 KW induction motor \mathcal{L}_2 optimal gain selection (left) vs explicit (Lspeed) selection (right) for the same α_{des}



Conclusions and perspectives

Conclusions

- Compact representation of the LTV dynamics of the IM
- Full-order Luenberger observer for the IM rotor flux estimation featuring
 - arbitrary global uniform exponential bounds on the estimation error, regardless of the rotor speed
 - ullet Optimal observer gains selection by \mathcal{L}_2 optimization
- Reduced-order observer covers existing results as special cases
- Simulation and experimental tests show the effectiveness of the proposed approach.

Future Work

- Follow the same approach for the dual control problem
- Can Kronecker-based "liftings" lead to novel ideas in motor control/estimation?

